

# AIP3N15A060Q4U Dual-In-Line Package Intelligent Power Module

# **External View**



Size: 38 x 24 x 3.6 mm



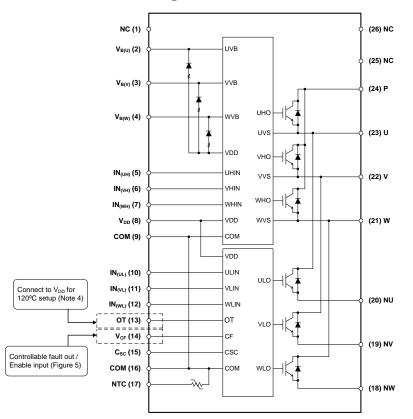
### Features

- 3-phase inverter module
- 600V-15A (Trench Shielded Planar Gate IGBT)
- · Built-in bootstrap diodes with integrated current-limiting resistor
- Control supply under-voltage lockout protection (UVLO)
- Controllable over-temperature protection (OT)
- Temperature monitoring (NTC)
- Short-circuit current protection (C<sub>SC</sub>)
- Controllable fault out signal (V $_{\rm CF}$ ) corresponding to SC, UV and OT fault
- Enable input functionality: Low-side IGBTs shut-down
- Input interface: 3 and 5V line, Schmitt trigger receiver circuit (Active high)
- Isolation ratings of 2000Vrms/min

### **Applications**

- AC 100-240Vrms class low power motor drives
- Air-conditioners, Washing machines, Compressors and Fan Motors

### Internal Equivalent Circuit / Pin Configuration





# **Ordering Information**

Part Number	Temperature Range	Package	Terminal type		
AIP3N15A060Q4U	-40°C to 150°C	IPM-3G	Ultra Short		
AOS Green Products use reduced levels of Halogens, and are also RoHS compliant. Please visit					

www.aosmd.com/media/AOSGreenPolicy.pdf for additional information.

# **Pin Description**

Green

Pin Number	Pin Name	Pin Function
1	NC	No Connection
2	V <sub>B(U)</sub>	High-Side Bias Voltage for U-Phase IGBT Driving
3	V <sub>B(V)</sub>	High-Side Bias Voltage for V-Phase IGBT Driving
4	V <sub>B(W)</sub>	High-Side Bias Voltage for W-Phase IGBT Driving
5	IN <sub>(UH)</sub>	Signal Input for High-Side U-Phase
6	IN <sub>(VH)</sub>	Signal Input for High-Side V-Phase
7	IN <sub>(WH)</sub>	Signal Input for High-Side W-Phase
8	V <sub>DD</sub>	Common Bias Voltage for IC and IGBTs Driving
9	COM	Common Supply Ground
10	IN <sub>(UL)</sub>	Signal Input for Low-Side U-Phase
11	IN <sub>(VL)</sub>	Signal Input for Low-Side V-Phase
12	IN <sub>(WL)</sub>	Signal Input for Low-Side W-Phase
13	ОТ	Controllable Over Temperature Protection (Connected to V <sub>DD</sub> for 120°C setup)
14	V <sub>CF</sub>	Controllable Fault Output
15	C <sub>SC</sub>	Capacitor (Low-Pass Filter) for Short-circuit Current Detection Input
16	COM	Common Supply Ground
17	NTC	Thermistor (Temperature Monitoring)
18	NW	Negative DC-Link Input for W-Phase
19	NV	Negative DC-Link Input for V-Phase
20	NU	Negative DC-Link Input for U-Phase
21	W	Output for W-Phase
22	V	Output for V-Phase
23	U	Output for U-Phase
24	Р	Positive DC-Link Input
25	NC	No Connection
26	NC	No Connection



# **Absolute Maximum Ratings**

 $T_J = 25^{\circ}C$ , unless otherwise specified.

Symbol	Parameter	Conditions	Ratings	Units
Inverter				
V <sub>PN</sub>	Supply Voltage	Applied between P - NU,NV,NW	450	V
V <sub>PN(surge)</sub>	Supply Voltage (surge)	Applied between P - NU,NV,NW	500	V
V <sub>CES</sub>	Collector-Emitter Voltage		600	V
	Output Dhoop Output	T <sub>C</sub> =25°C, T <sub>J</sub> <150°C	15	А
l <sub>C</sub>	Output Phase Current	T <sub>C</sub> =80°C, T <sub>J</sub> <150°C	10	А
±I <sub>PK</sub>	Output Peak Phase Current	T <sub>C</sub> =25°C, less than 1ms pulse width	30	А
t <sub>SC</sub>	Short Circuit Withstand Time	V <sub>PN</sub> ≤400V, T <sub>J</sub> =150°C, V <sub>DD</sub> =15V	5	μs
Pc	Collector Dissipation	T <sub>C</sub> =25°C, per chip	33	W
TJ	Operating Junction Temperature		-40 to 150	°C
Control (P	rotection)			
V <sub>DD</sub>	Control Supply Voltage	Applied between V <sub>DD</sub> -COM	25	V
V <sub>DB</sub>	High-Side Control Bias Voltage	Applied between V <sub>B(U)</sub> -U, V <sub>B(V)</sub> -V, V <sub>B(W)</sub> -W	25	V
V <sub>IN</sub>	Input Voltage	$\begin{array}{l} \mbox{Applied between } IN_{(UH),} \ IN_{(VH),} \ IN_{(WH),} \ IN_{(UL),} \\ IN_{(VL),} \ IN_{(WL)} - COM \end{array}$	V <sub>DD</sub> +0.3	V
$V_{CF}$	Fault Output Supply Voltage	Applied between V <sub>CF</sub> -COM	COM+5.5	V
I <sub>CF</sub>	Fault Output Current	Sink current at V <sub>CF</sub> terminal	1	mA
Vsc	Current Sensing Input Voltage	Applied between C <sub>SC</sub> -COM	COM+5.5	V
Total Syst	em	· · · · · · · · · · · · · · · · · · ·		
V <sub>PN(PROT)</sub>	Self Protection Supply Voltage Limit (Short-Circuit Protection Capability)	$V_{DD}$ =13.5-16.5V, Inverter part $T_J$ =150°C, Non-repetitive, less than 2µs	400	V
T <sub>c</sub>	Module Case Operation Temperature	Measurement point of T <sub>C</sub> is provided in Figure 1	-30 to 125	°C
T <sub>STG</sub>	Storage Temperature		-40 to 150	°C
V <sub>ISO</sub>	Isolation Voltage	60Hz, sinusoidal, AC 1min, between connected all pins and heat sink plate	2000	V <sub>rms</sub>

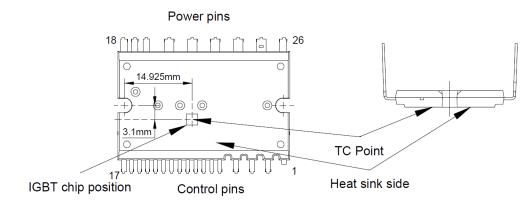


Figure 1. T<sub>C</sub> Measurement Point



# **Thermal Resistance**

Symbol	Parameter	Conditions	Min.	Тур.	Max.	Units
R <sub>th(j-c)Q</sub>		Inverter IGBT (per 1/6 module)	-	-	3.8	K/W
R <sub>th(j-c)F</sub>	Junction to Case Thermal Resistance <sup>(1)</sup>	Inverter FWD (per 1/6 module)	-	-	4.97	K/W

Note:

1. For the measurement point of case temperature ( $T_c$ ), please refer to Figure 1.

### **Electrical Characteristics**

 $T_J = 25^{\circ}C$ , unless otherwise specified.

Symbol	Parameter	Co	nditions	Min.	Тур.	Max.	Units
Inverter	•	•					
M	Collector-Emitter Saturation	V <sub>DD</sub> =V <sub>DB</sub> =15V,	I <sub>C</sub> =7.5A, T <sub>J</sub> =25°C	-	1.40	1.90	V
$V_{CE(SAT)}$	Voltage	V <sub>IN</sub> =5V	I <sub>C=7.5</sub> A, T <sub>J</sub> =125°C	-	1.60	-	V
V <sub>F</sub>	FWD Forward Voltage	V <sub>IN</sub> =0	I <sub>F</sub> =7.5A, T <sub>J</sub> =25°C	-	1.55	2.00	V
t <sub>ON</sub>				0.40	0.70	1.20	μs
t <sub>C(ON)</sub>		V <sub>PN</sub> =300V, V <sub>DD</sub> =V <sub>DB</sub> =	=15V	-	0.15	0.40	μs
t <sub>OFF</sub>		$I_{C}$ =10A, $T_{J}$ =25°C, $V_{IN}$	$\downarrow$ =0V $\leftrightarrow$ 5V	-	1.25	1.75	μs
$t_{C(OFF)}$		Inductive load		-	0.10	0.30	μs
t <sub>rr</sub>				-	0.10	-	μs
	Collector-Emitter Leakage		T <sub>J</sub> =25°C	-	-	1	mA
ICES	Current	V <sub>CE</sub> =V <sub>CES</sub>	T <sub>J</sub> =125°C	-	-	10	mA
Control (P	rotection)						
I <sub>QDD</sub>	Quiescent V <sub>DD</sub> Supply Current	V <sub>DD</sub> =15V, IN <sub>(UH,VH,WH,UL,VL,WL)</sub> =0V	V <sub>DD</sub> -COM	-	-	2.1	mA
I <sub>QDB</sub>	Quiescent V <sub>DB</sub> Supply Current	V <sub>DB</sub> =15V, IN <sub>(UH, VH, WH)</sub> =0V	$V_{B(U)}\text{-}U,  V_{B(V)}\text{-}V,  V_{B(W)}\text{-}W$	-	-	0.3	mA
V <sub>SC(ref)</sub>	Short-Circuit Trip Level	V <sub>DD</sub> =15V <sup>(2)</sup>		0.455	0.48	0.505	V
UV <sub>DT</sub>		Trip Level		10.3	11.4	12.5	V
UV <sub>DR</sub>	Supply Circuit Under-Voltage	Reset Level		10.8	11.9	13.0	V
UV <sub>DBT</sub>	Protection	Trip Level		8.5	9.5	10.5	V
UV <sub>DBR</sub>		Reset Level		9.5	10.5	11.5	V
OTT	Over-Temperature	The OT Pin is	Trip Level	100	120	140	°C
OT <sub>HYS</sub>	Protection <sup>(3)</sup>	connected to V <sub>DD</sub> or open	Hysteresis of Trip Reset	-	30	-	°C



Symbol	Parameter	Conditions	Min.	Тур.	Max.	Units
Control (P	Protection)	•				
$V_{CFH}$		$V_{SC}$ =0V, $V_{CF}$ Circuit: 10k $\Omega$ to 5V pull-up	4.9	-	-	V
$V_{CFL}$	Fault Output Voltage	$V_{SC}$ =1V, $V_{CF}$ Circuit: 10k $\Omega$ to 5V pull-up	-	-	0.5	V
$V_{CF+}$	CF positive going threshold		-	1.9	2.2	V
V <sub>CF-</sub>	CF negative going threshold		0.8	1.1	-	V
		Pull-up resistor only	20	-	-	μs
t <sub>FO</sub>	Fault Output Pulse Width <sup>(4)</sup>	Pull-up resistor with pull-down capacitor $(R_{CF}=2.2M\Omega, C_{CF}=1nF, 5V \text{ pull-up})^{(Figure 5)}$	-	1	-	ms
l <sub>in</sub>	Input Current	V <sub>IN</sub> =5V	-	0.72	-	mA
V <sub>th(on)</sub>	ON Threshold Voltage			2.3	2.6	V
V <sub>th(off)</sub>	OFF Threshold Voltage	Applied between $IN_{(UH)}$ , $IN_{(VH)}$ , $IN_{(WH)}$ , $IN_{(UL)}$ ,	0.8	1.2		V
$V_{\text{th(hys)}}$	ON/OFF Threshold Hysteresis Voltage	IN <sub>(VL)</sub> , IN <sub>(WL)</sub> -COM	-	1.1	-	V
$V_{F(BSD)}$	Bootstrap Diode Forward Voltage	I <sub>F</sub> =10mA Including Voltage Drop by Limiting Resistor <sup>(5)</sup>	0.5	1.0	1.5	V
R <sub>BSD</sub>	Built-in Limiting Resistance	Included in Bootstrap Diode	80	100	120	Ω

#### Notes:

2. Short-circuit protection works only for low sides.

3. When the LVIC temperature exceeds OT Trip temperature level (OT<sub>T</sub>), OT protection is triggered and fault outputs. OT Trip level can be adjusted by pull-down resistors values as shown in the table below.

OT Pin	OT <sub>⊺</sub> [°C]
10kΩ	Disable
100kΩ	130
400kΩ	110
$V_{\text{DD}}$ or Open	120

- 4. Fault signal (F<sub>o</sub>) outputs when SC, UV or OT protection is triggered. F<sub>o</sub> pulse width is different for each protection mode. At SC failure, F<sub>o</sub> pulse width is fixed (minimum 20µs) or controlled by RC network (see Figure 5), but at UV or OT failure, F<sub>o</sub> outputs continuously until recovering from UV or OT state.
- 5. The characteristics of bootstrap diodes are shown in Figure 2.

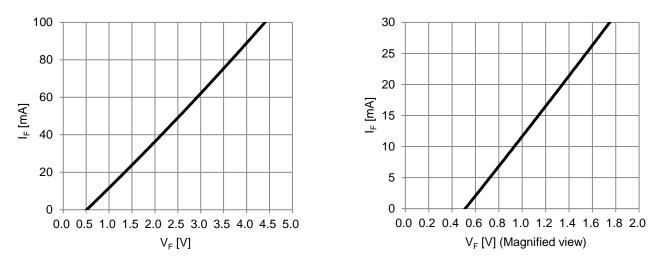


Figure 2. Built-in Bootstrap Diode V<sub>F</sub>-I<sub>F</sub> Characteristic (T<sub>C</sub>=25°C)



### **NTC Thermistor**

Symbol	Parameter	Conditions	Min.	Тур.	Max.	Units
R <sub>25</sub>	Resistance	T <sub>NTC</sub> =25°C	-	84.83	-	kΩ
B(25/100)	C-constant		-	4092	-	К

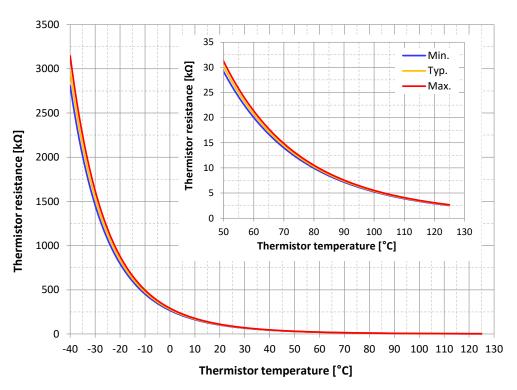
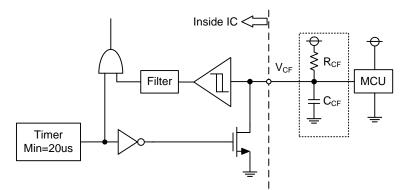


Figure 3. Thermistor Resistance vs. Temperature Curve



- (1) The V<sub>CF</sub> pin combines three functions in one pin: Fixed fault out, Controllable fault out pulse width based on RC network, and Enable input.
- (2) The  $V_{CF}$  pin provides an enable functionality that allows it to shut down the all low-side IGBTs. When the  $V_{CF}$  pin is in the high state the IPM is able to operate normally. If the  $V_{CF}$  pin is in a low state, the low-side IGBTs are turned off until the enable condition is restored. In addition, the  $V_{CF}$  pin can provide the fault output signal with the fixed or controlled fault out pulse width.
- (3) If a pull-up resistor ( $10k\Omega$ ) only is connected to the V<sub>CF</sub> pin, the fault output pulse width is fixed at minimum 20us.
- (4) If a capacitor (C<sub>CF</sub>) is connected with a pull-up resistor (R<sub>CF</sub>) together, the fault output pulse width can be controlled according to the resistor and the capacitor values. The length of fault output pulse width is determined by the following formula;
  - $t_{FO} = -(R_{CF}*C_{CF})*ln(1-V_{CF}+/V_{DD}) + 100ns + 20us(min.)$
  - ex)  $V_{DD}=5V$ ,  $R_{CF}=2.2M\Omega$ ,  $C_{CF}=1nF$ ,  $t_{FO}\approx1.07ms$ . Recommended parameters in the design are  $C_{CF}$  of  $\leq 1nF$  and  $R_{CF}$  of 0.1M to 2.2M $\Omega$ .

Figure 4. Interface Circuit at Pin VCF

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# **Mechanical Characteristics and Ratings**

Symbol	Parameter	Conditions	Min.	Тур.	Max.	Units
Mounting Torque	Mounting Screw: M3 <sup>(6)</sup>		0.59	0.69	0.78	N m
Weight			-	9.12	-	g
Flatness	Refer to Figure 6		-50	-	100	μm

#### Note:

6. Plain washers (ISO 7089-7094) are recommended.

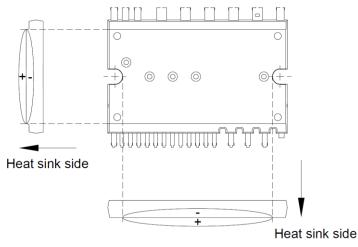


Figure 5. Flatness Measurement Position

## **Recommended Operation Conditions**

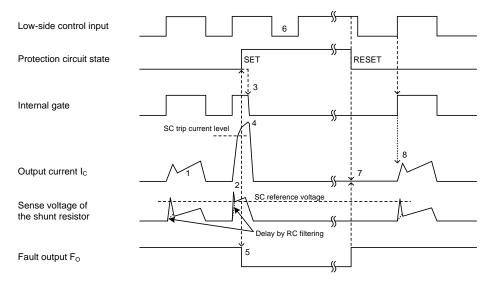
Symbol	Parameter	Conditions	Min.	Тур.	Max.	Units
V <sub>PN</sub>	Supply Voltage	Applied between P-NU, NV, NW	0	300	400	V
V <sub>DD</sub>	Control Supply Voltage	Applied between VDD-COM	13.5	15.0	16.5	V
V <sub>DB</sub>	High-Side Bias Voltage	Applied between $V_{B(U)}$ -U, $V_{B(V)}$ -V, $V_{B(W)}$ -W	13.5	15.0	18.5	V
dV <sub>DD</sub> /dt, dV <sub>DB</sub> /dt	Control Supply Variation		-1	-	+1	V/µs
t <sub>dead</sub>	Arm Shoot-Through Blocking Time	For each input signal	1.5	-	-	μs
<b>f</b> <sub>PWM</sub>	PWM Input Frequency	-40°C < T <sub>J</sub> < 150°C	-	-	20	kHz
PW <sub>IN(ON)</sub>			0.5	-	-	μs
$PW_{IN(OFF)}$	Minimum Input Pulse Width <sup>(7)</sup>		0.5	-	-	μs
СОМ	COM Variation	Between COM-NU, NV, NW (including surge)	-5.0	-	5.0	V

#### Note:

7. IPM may not respond if the input pulse width is less than  $\text{PW}_{\text{IN(ON)}}, \text{PW}_{\text{IN(OFF)}}.$ 



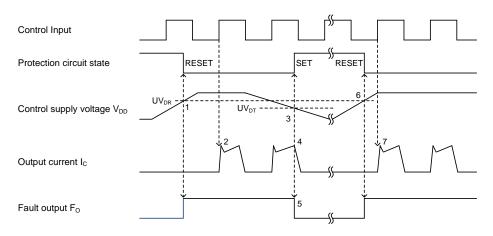
### **Time Charts of the IPM Protective Function**



(1) Normal operation: IGBT turns on and outputs current.

- (2) Short-circuit current detection (SC triggered).
- (3) All low-side IGBTs' gates are hard interrupted.
- (4) All low-side IGBTs turn OFF.
- (5)  $F_{O}$  output time (t<sub>FO</sub>)=minimum 20µs.
- (6) Input = "L" : IGBT OFF.
- (7) Fault output finishes, but output current will not turn on until next ON signal (L $\rightarrow$ H).
- (8) Normal operation: IGBT turns on and outputs current.

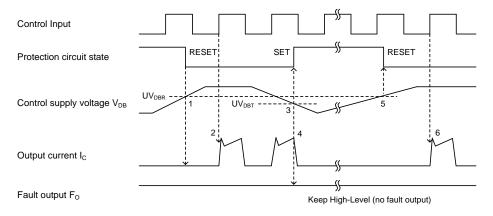
### Figure 6. Short-Circuit Protection (Low-side Operation Only with the External Shunt Resistor and RC Filter)



- (1) Control supply voltage  $V_{DD}$  exceeds under voltage reset level (UV<sub>DR</sub>), but IGBT turns on by next ON signal (L $\rightarrow$ H).
- (2) Normal operation: IGBT turns on and outputs current.
- (3)  $V_{DD}$  level drops to under voltage trip level (UV<sub>DT</sub>).
- (4) All low-side IGBTs turn OFF regardless of control input condition.
- (5)  $F_O$  output time (t<sub>FO</sub>)=minimum 20µs, and  $F_O$  stays low as long as  $V_{DD}$  is below UV<sub>DR</sub>.
- (6)  $V_{DD}$  level reaches  $UV_{DR}$ .
- (7) Normal operation: IGBT turns on and outputs current.

#### Figure 7. Under-Voltage Protection (Low-side, UV<sub>D</sub>)





(1) Control supply voltage  $V_{DB}$  rises. After the voltage reaches under voltage reset level UV<sub>DBR</sub>, IGBT turns on by next ON signal (L $\rightarrow$ H).

(2) Normal operation: IGBT turns on and outputs current.

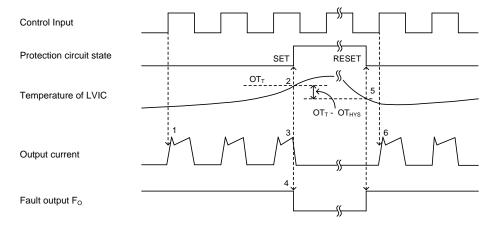
(3) V<sub>DB</sub> level drops to under voltage trip level (UV<sub>DBT</sub>).

(4) All high-side IGBTs turn OFF regardless of control input condition, but there is no Fo signal output.

(5) V<sub>DB</sub> level reaches UV<sub>DBR</sub>.

(6) Normal operation: IGBT turns on and outputs current.

#### Figure 8. Under-Voltage Protection (High-side, UV<sub>DB</sub>)



(1) Normal operation: IGBT turns on and outputs current.

(2) LVIC temperature exceeds over-temperature trip level  $(OT_T)$ .

(3) All low-side IGBTs turn off regardless of control input condition.

(4)  $F_0$  output time (t<sub>F0</sub>)=minimum 20µs, and  $F_0$  stays low as long as LVIC temperature is over OT<sub>T</sub>.

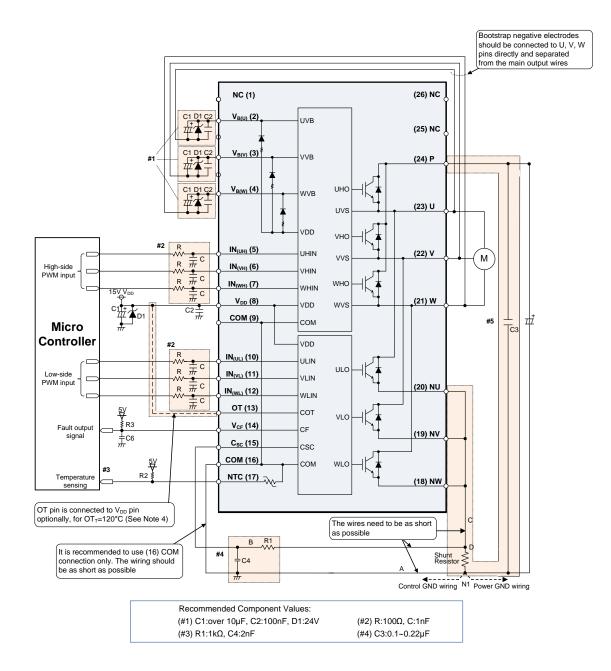
(5) LVIC temperature drops to over-temperature reset level (OT<sub>T</sub>-OT<sub>HYS</sub>).

(6) Normal operation: IGBT turns on by the next ON signal ( $L \rightarrow H$ ).

#### Figure 9. Over-Temperature Protection (Low-side, Detecting LVIC Temperature)

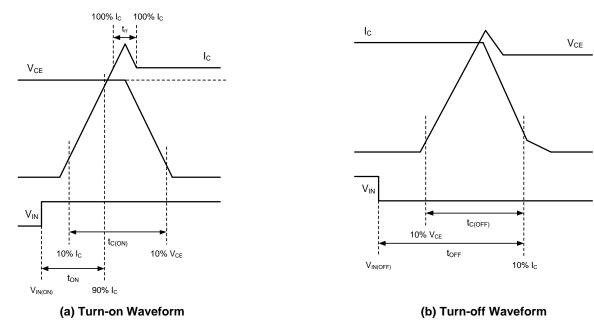


### **Example of Application Circuit**



- (1) GND pattern: It is recommended to connect the control GND and power GND at a single point (N1). GND pattern should be separated at the one point of the shunt resistors.
- (2) COM pin: It is recommended to only use the (16) COM pin to minimize SC detection noise. Leave pin (9) NC (No Connection).
- (3) A Zener diode D1 (24V/1W) is recommended between each pair of control supply pins to prevent surge destruction.
- (4) Snubber capacitor: The wiring between the IPM and snubber capacitor (C3) including the shunt resistors should be as short as possible.
- (5) C<sub>SC</sub> pin circuit: C4 should be placed as close to C<sub>SC</sub> pin and COM (16) pin as possible to prevent protection function errors.
- (6) Bootstrap capacitors: It is recommended that all capacitors are mounted as close to the IPM as possible.
- (7) Input circuit: The R and C filter circuit should be mounted to reduce input signal noise by high speed switching. C should be placed as close to COM (16) pin as possible.
- (8)  $V_{CF}$  pin circuit:  $V_{CF}$  output is open drain type. The signal line should be pulled up to the positive side of the 5V/3.3V logic power supply with a proper resistor R3. For the detailed design guide, please refer to the Figure 5.



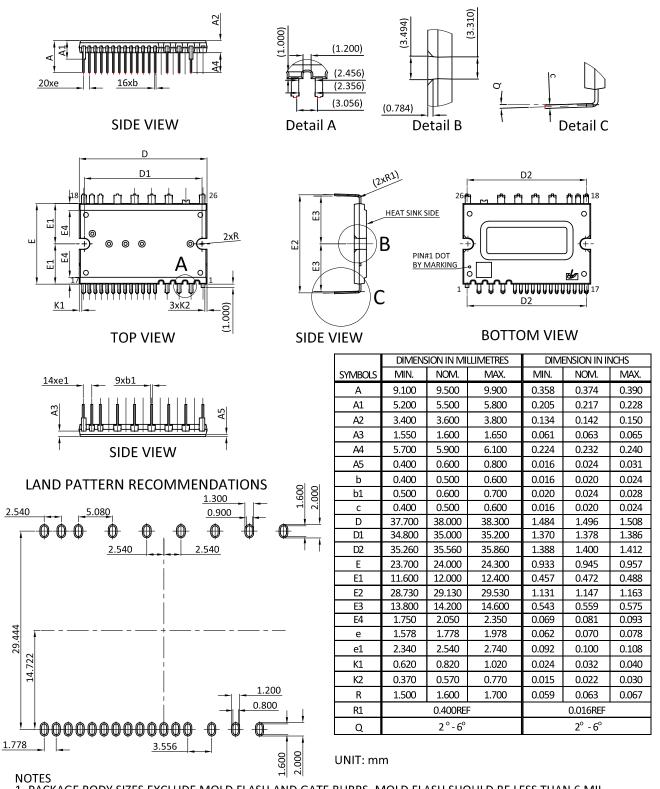






## **Package Dimensions**

IPM-3G: Ultra Short Terminal Type



1. PACKAGE BODY SIZES EXCLUDE MOLD FLASH AND GATE BURRS, MOLD FLASH SHOULD BE LESS THAN 6 MIL.

2. TOLERANCE 0.100 MILLIMETERS UNLESS OTHERWISE SPECIFIED.

3. CONTROLLING DIMENSION IS MILLIMETER, CONVERTED INCH DIMENSIONS ARE NOT NECESSARILY EXACT.

4. () IS REFERENCE.



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